# LMX2542 PLLatinum ${ }^{\text {TM }}$ Cellular and GPS Frequency Synthesizer System with Integrated vCO 

Check for Samples: LMX2542

## FEATURES

- Small Size
- $5.0 \mathrm{~mm} \times 5.0 \mathrm{~mm} \times 0.75 \mathrm{~mm}$ 28-Pin WQFN
- RF Synthesizer System
- Integrated RF VCO
- Integrated Loop Filter
- Low Spurious, Low Phase Noise FractionalN RF PLL Based on 11-Bit $\Delta \Sigma$ Modulator
- 5 kHz Frequency Resolution
- Cellular-CDMA LO: 2105.28 MHz to 2155.14 MHz (Requires an External LO /2 Circuit)
- GPS LO: 2087.73 MHz
- (Requires an External LO /1.5 Circuit)
- IF Synthesizer System
- Integer-N IF PLL
- Programmable Charge Pump Current Levels
- IF LO: $\mathbf{3 6 7 . 2 0} \mathbf{M H z}$
- Supports Various Reference Oscillator Frequencies:
- 19.20 MHz 19.68 MHz
- Low Current Consumption:
- 22 mA Typical at 2.8 V
- 2.7V to 3.3V Operation
- RF Digital Filtered Lock Detect Output
- Hardware and Software Powerdown Control


## DESCRIPTION

LMX2542 is a highly integrated, high performance, low power frequency synthesizer system optimized for Cellular-CDMA 1xRTT and IS-95 mobile handsets and data systems with GPS capabilities. Using a proprietary digital phase locked loop technique, LMX2542 provides very stable, low noise local oscillator (LO) signals for up and down conversion in wireless communications devices.
LMX2542 includes a Voltage Controlled Oscillator (VCO) for both the Cellular-CDMA and GPS frequency bands, a loop filter, and a Fractional-N RF PLL based on a Delta Sigma ( $\Delta \Sigma$ ) modulator. In concert, these blocks form a closed loop RF synthesizer system. The RF synthesizer system operates from 2087.73 MHz to 2155.14 MHz .
LMX2542 includes an Integer-N IF PLL also. For more flexible loop filter designs, the IF PLL includes a 4-level programmable charge pump. Together with an external VCO and loop filter, LMX2542 makes a complete closed loop IF synthesizer system. The default IF frequency is 367.20 MHz .
Serial data is transferred to the device via a threewire MICROWIRE ${ }^{\text {TM }}$ interface (DATA, LE, CLK).
Operating supply voltage ranges from 2.7 V to 3.3 V . LMX2542 features low current consumption: 22 mA at 2.8 V .

LMX2542 is available in a 28 -Pin Leadless Leadframe Package (WQFN).

## APPLICATIONS

- Cellular-CDMA 1xRTT and IS-95 Mobile Handsets with GPS
- Cellular-CDMA 1xRTT and IS-95 Mobile Data Systems with GPS

Leadless Leadframe Package


[^0]
## Functional Block Diagram



## Connection Diagram



Note: Analog GND connected through exposed die attached pad.

## Leadless Leadframe Package (WQFN) (Top View) <br> See Package Number NJB

PIN DESCRIPTION

| Pin No. | Pin Name | I/O |  |
| :---: | :---: | :---: | :--- |
| 1 | Fin | I | IF PLL buffer/prescaler input. Small signal input from the VCO. |
| 2 | V | VC | - |
| Power supply bias for the IF PLL analog circuits. VCC may range from 2.7V to 3.3V. Bypass <br> capacitors should be placed as close as possible to this pin and be connected directly to the <br> ground plane on the printed circuit board. |  |  |  |
| 3 | CPout | O | IF PLL charge pump output. The output is connected to the external loop filter, which drives <br> the input of the IF VCO. |
| 4 | NC | - | No Connect. Do not connect to any node on the printed circuit board. |
| 5 | LE | I | MICROWIRE Latch Enable Input. High impedance CMOS input. When LE transitions from <br> LOW to HIGH, DATA stored in the shift register is loaded into one of 6 internal control <br> registers. |
| 6 | CLK | I | MICROWIRE Clock Input. High impedance CMOS input. DATA is clocked into the 24-bit shift <br> register on the rising edge of CLK. |
| 7 | DATA | I | MICROWIRE Data Input. High impedance CMOS input. Binary serial data. The MSB of <br> DATA is shifted in first. |
| 9 | VDD | - | Powe supply bias for the RF VCO. V VD may range from 2.7V to 3.3V. Bypass capacitors <br> should be placed as close as possible to this pin and be connected directly to the ground <br> plane on the printed circuit board. |
| 10 | NC | - | No Connect. Do not connect to any node on the printed circuit board. |
| 11 | NC | - | No Connect. Do not connect to any node on the printed circuit board. |
| 12 | NC | - | No Connect. Do not connect to any node on the printed circuit board. |

PIN DESCRIPTION (continued)

| Pin No. | Pin Name | 1/0 | Description |
| :---: | :---: | :---: | :---: |
| 13 | $V_{D D}$ | - | Power supply bias for the RF VCO. $\mathrm{V}_{\mathrm{DD}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 14 | $V_{D D}$ | - | Power supply bias for the RF VCO output buffer. $\mathrm{V}_{\mathrm{DD}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 15 | RFout | O | Buffered RF VCO output. |
| 16 | $\mathrm{V}_{\mathrm{Cc}}$ | - | Power supply bias for the RF PLL prescaler. $\mathrm{V}_{\mathrm{CC}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 17 | $\mathrm{V}_{\mathrm{CC}}$ | - | Power supply bias for the RF PLL charge pump. $\mathrm{V}_{C C}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 18 | $\mathrm{V}_{\mathrm{CC}}$ | - | Power supply bias for the RF PLL digital circuits. $\mathrm{V}_{\mathrm{CC}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 19 | LD | O | Digital filtered lock detect output. |
| 20 | CE | I | Chip Enable input. High Impedance CMOS input. When this pin is set HIGH, the RF and IF synthesizer systems are powered up. Powerdown is then controlled through the MICROWIRE. When this pin is set LOW, the device is asynchronously powered down and the IF PLL charge pump output is forced to a high impedance state (Tri-State). |
| 21 | GND | - | Ground for the RF PLL digital circuits. |
| 22 | OSCin | I | Reference oscillator input. The input is driven by an external AC coupled source. When the OSC_FREQ bit is set LOW, a 19.20 MHz reference frequency should be used. When the OSC_FREQ bit is set HIGH, a 19.68 MHz reference frequency should be used. |
| 23 | $\mathrm{V}_{\mathrm{Cc}}$ | - | Power supply bias for the reference oscillator buffer. $\mathrm{V}_{\mathrm{CC}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 24 | GND | - | Ground for the reference oscillator buffer. |
| 25 | GND | - | Ground for the IF PLL digital circuits. |
| 26 | $\mathrm{V}_{\mathrm{CC}}$ | - | Power supply bias for the IF PLL digital circuits. $\mathrm{V}_{\mathrm{CC}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |
| 27 | NC | - | No Connect. Do not connect to any node on the printed circuit board. |
| 28 | $\mathrm{V}_{\mathrm{CC}}$ | - | Power supply bias for the IF PLL buffer/ prescaler. $\mathrm{V}_{\mathrm{CC}}$ may range from 2.7 V to 3.3 V . Bypass capacitors should be placed as close as possible to this pin and be connected directly to the ground plane on the printed circuit board. |

## Part Number Description



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## Absolute Maximum Ratings ${ }^{(1)(2)(3)(4)}$

| Power Supply Voltage | $\mathrm{V}_{\text {CC }}$ to GND | -0.3 V to +3.6 V |
| :---: | :---: | :---: |
|  | $V_{\text {DD }}$ to GND | -0.3 V to +3.6 V |
| Voltage on any pin to GND ( $\mathrm{V}_{\text {IN }}$ ) | $\mathrm{V}_{\text {IN }}$ must be $<+3.6 \mathrm{~V}$ | -0.3 V to $\mathrm{V}_{\mathrm{Cc}}+0.3 \mathrm{~V}$ |
|  |  | -0.3 V to $\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}$ |
| Storage Temperature Range ( $\mathrm{T}_{\mathrm{S}}$ ) |  | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Lead Temperature (solder 4 s ) ( $\mathrm{T}_{\mathrm{L}}$ ) |  | $+260^{\circ} \mathrm{C}$ |

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Recommended Operating Conditions indicate conditions for which the device is intended to be functional, but do not ensure specific performance limits. For ensured specifications and test conditions, refer to the Electrical Characteristics section. The ensured specifications apply only for the conditions listed.
(2) This device is a high performance RF integrated circuit with an ESD rating $<2 k V$ and is ESD sensitive. Handling and assembly of this device should be done at ESD protected work stations.
(3) $\mathrm{GND}=0 \mathrm{~V}$.
(4) If Military/Aerospace specified devices are required, please contact the TI Sales Office/ Distributors for availability and specifications.

## Recommended Operating Conditions

| Power Supply Voltage | $\mathrm{V}_{\mathrm{CC}}$ to GND | +2.7 V to +3.3 V |
| :--- | :--- | ---: |
|  | $\mathrm{~V}_{\mathrm{DD}}$ to GND | +2.7 V to +3.3 V |
| Operating Temperature $\left(\mathrm{T}_{\mathrm{A}}\right)$ | $-30^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  |

## Electrical Characteristics

$V_{C C}=V_{D D}=C E=2.8 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise specified

| Symbol | Parameter | Conditions | Value |  |  | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max |  |
| Icc PARAMETERS |  |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{CC}}+\mathrm{I}_{\mathrm{DD}}$ | Power Supply Current <br> (RF and IF Synthesizer Systems) | RF_EN Bit = 1 <br> IF_EN Bit = 1 <br> OB_CRL[1:0] Word $=00$ <br> VCO_CUR[1:0] Word = 11 <br> IF_CUR[1:0] Word $=00$ |  | 22.0 | 24.0 | mA |
| (lcc + $\left.l_{\mathrm{DD}}\right)_{\mathrm{RF}}$ | Power Supply Current (RF Synthesizer System) | ```RF_EN Bit = 1 IF_EN Bit = 0 O\overline{B}_CRL[1:0] Word = 00 VCO_CUR[1:0] Word = 11``` |  | 20.0 | 22.0 | mA |
| $\mathrm{I}_{\text {PD }}$ | Powerdown Current | $\begin{aligned} & \text { CE, CLK, DATA and LE = OV } \\ & \text { OSCin = OV } \\ & (\text { RF_EN Bit }=0 \text { and IF_EN Bit }=0) \end{aligned}$ |  |  | 20.0 | $\mu \mathrm{A}$ |
| RF SYNTHESIZER SYSTEM PARAMETERS |  |  |  |  |  |  |
| RF VCO |  |  |  |  |  |  |
| $\mathrm{f}_{\text {RFout }}$ | RF VCO Operating Frequency ${ }^{(1)(2)}$ |  | 2087.73 |  | 2155.14 | MHz |
| PRFout | RF VCO Output Power | OB_CRL[1:0] Word $=00$ | -7.5 | -4.5 | -1.5 | dBm |
|  |  | OB_CRL[1:0] Word $=01$ | -5.0 | -2.0 | 1.0 | dBm |
|  |  | OB_CRL[1:0] Word $=10$ | -2.5 | 0.5 | 3.5 | dBm |
|  |  | OB_CRL[1:0] Word = 11 | 0.0 | 3.0 | 6.0 | dBm |
| $\varphi_{\text {eRF }}$ | RF VCO RMS Phase Error |  |  | 1.3 |  | Deg. |
| $L_{\text {RF }}(\mathrm{f})$ | RF VCO Single Side Band Phase Noise | $\mathrm{f}=100 \mathrm{kHz}$ Offset TCXO Reference Source OSC_FREQ Bit $=0$ or 1 OB_CRL[1:0] Word = 11 IF_EN Bit = 0 |  | -109 | -107 | $\begin{gathered} \mathrm{dBc} / \\ \mathrm{Hz} \end{gathered}$ |
|  |  | $\mathrm{f}=900 \mathrm{kHz}$ Offset TCXO Reference Source OSC_FREQ Bit $=0$ or 1 OB_CRL[1:0] Word = 11 IF_EN Bit = 0 |  | -134 | -133 | $\begin{gathered} \mathrm{dBc} / \\ \mathrm{Hz} \end{gathered}$ |
| SPURS ${ }_{\text {RF }}$ | RF Synthesizer Reference Spurs | OSC_FREQ Bit $=0$ or 1 IF_EN Bit = 0 |  |  | -75 | dBc |
| $\mathrm{HS}_{\mathrm{RF}}$ | RF VCO Harmonic Suppression | $\begin{aligned} & 2^{\text {ND }} \text { Harmonic } \\ & \text { OB_CRL[1:0] Word = } 11 \end{aligned}$ |  |  | -25 | dBc |
|  |  | $\begin{aligned} & 3^{\mathrm{RD}} \text { Harmonic } \\ & \text { OB_CRL[1:0] Word }=11 \end{aligned}$ |  |  | -25 | dBc |
| $t_{\text {RFLOCK }}$ | Channel Switch Lock Time ${ }^{(3)}$ | $\begin{aligned} & \mathrm{f}_{\text {INITIAL }}=2087.73 \mathrm{MHz} \\ & \mathrm{f}_{\text {FINAL }}=2155.14 \mathrm{MHz} \end{aligned}$ |  | 1.0 | 1.3 | ms |

(1) For other RF frequency ranges, please contact Texas Instruments.
(2) When the Cellular-CDMA mode is used, an external /2 circuit is required before the Cellular mixer LO port. Furthermore, if an external $/ 1.5$ circuit is available before the GPS mixer LO port, the GPS frequency of 1391.82 MHz can be achieved by using a fixed RF frequency of 2087.73 MHz .
(3) Lock time is defined as the time difference between the beginning of the frequency transition and the point at which the frequency remains within $+/-1 \mathrm{kHz}$ of the final frequency. $\mathrm{t}_{\text {LOCK }}=\mathrm{t}_{\text {FINAL }}-\mathrm{t}_{\text {INITIAL }}$.

## Electrical Characteristics (continued)

$\mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{DD}}=\mathrm{CE}=2.8 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise specified

| Symbol | Parameter | Conditions | Value |  |  | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max |  |
| IF SYNTHESIZER SYSTEM PARAMETERS |  |  |  |  |  |  |
| $\mathrm{f}_{\text {Fin }}$ | IF Synthesizer Operating Frequency ${ }^{(4)}$ | $\begin{aligned} & \text { SPI_DEF Bit = } 1 \\ & \text { IF_FREQ[1:0] Word = } 00 \\ & \hline \end{aligned}$ |  | 170.76 |  | MHz |
|  |  | $\begin{aligned} & \text { SPI_DEF Bit = } 1 \\ & \text { IF_FREQ[1:0] Word = } 01 \\ & \text { (Default) } \end{aligned}$ |  | 367.20 |  | MHz |
|  |  | $\begin{aligned} & \text { SPI_DEF Bit = } 1 \\ & \text { IF_FREQ[1:0] Word = } 10 \end{aligned}$ |  | 440.76 |  | MHz |
| $\mathrm{f}_{\varphi \mathrm{lF}}$ | IF Synthesizer Phase Detector Frequency |  |  | 120 |  | kHz |
| $\mathrm{p}_{\text {Fin }}$ | IF Synthesizer Input Sensitivity |  | -12 |  | 0 | dBm |
| $\mathrm{I}_{\text {CPoutIF }}$ | IF Synthesizer Charge Pump Output Current | IF_CUR[1:0] Word = 00 |  | 100 |  | $\mu \mathrm{A}$ |
|  |  | IF_CUR[1:0] Word $=01$ |  | 200 |  | $\mu \mathrm{A}$ |
|  |  | IF_CUR[1:0] Word $=10$ |  | 300 |  | $\mu \mathrm{A}$ |
|  |  | IF_CUR[1:0] Word = 11 |  | 800 |  | $\mu \mathrm{A}$ |

## REFERENCE OSCILLATOR PARAMETERS

| $\mathrm{f}_{\text {OSCin }}$ | Reference Oscillator Input Operating <br> Frequency | OSC_FREQ Bit $=0$ | 19.20 |  |  | MHz |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
|  | OSC_FREQ Bit $=1$ |  |  | 19.68 | MHz |  |
| $\mathrm{V}_{\text {OSCin }}$ | Reference Oscillator Input Sensitivity |  |  | 0.2 | $\mathrm{~V}_{\mathrm{CC}}$ | $\mathrm{V}_{\text {PP }}$ |

DIGITAL INTERFACE (CE, DATA, CLK, LE, LD)

| $\mathrm{V}_{\mathrm{IH}}$ | High-Level Input Voltage |  | $0.8 \mathrm{~V}_{\mathrm{DD}}$ |  | $\mathrm{V}_{\mathrm{DD}}$ | V |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  | $0.8 \mathrm{~V}_{\mathrm{CC}}$ |  | $\mathrm{V}_{\mathrm{CC}}$ | V |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-Level Input Voltage |  | 0 |  | $0.2 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  |  |  | 0 |  | $0.2 \mathrm{~V}_{\mathrm{CC}}$ | V |
| $\mathrm{I}_{\mathrm{IH}}$ | High-Level Input Current | $\mathrm{V}_{\mathrm{IH}}=\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{CC}}$ |  |  | 10 | $\mu \mathrm{~A}$ |
| $\mathrm{I}_{\mathrm{IL}}$ | Low-Level Input Current | $\mathrm{V}_{\mathrm{IL}}=0 \mathrm{~V}$ | -10 |  |  |  |
| $\mathrm{C}_{\mathrm{I}}$ | Input Capacitance |  |  | 3.0 |  |  |
| $\mathrm{~V}_{\mathrm{OH}}$ | High-Level Output Voltage |  | $0.9 \mathrm{~V}_{\mathrm{DD}}$ | pF |  |  |
| $\mathrm{V}_{\mathrm{OL}}$ | Low-Level Output Voltage | $0.9 \mathrm{~V}_{\mathrm{CC}}$ |  |  | V |  |
| $\mathrm{C}_{\mathrm{O}}$ | Output Capacitance |  |  |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |

(4) For frequencies other than the default values, the SPI_DEF bit should be set to 0 and registers R4 and R5 programmed appropriately. Refer to Serial Port Interface Default Register Selection for further details on how to program the SPI_DEF bit.
(5) For other reference oscillator frequencies, please contact Texas Instruments.

## Electrical Characteristics (continued)

| Symbol | Parameter | Conditions | Value |  |  | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max |  |
| MICROWIRE INTERFACE |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{CS}}$ | DATA to CLK Set Up Time |  | 50.0 |  |  | ns |
| $\mathrm{t}_{\mathrm{CH}}$ | DATA to CLK Hold Time |  | 10.0 |  |  | ns |
| $\mathrm{t}_{\text {cWh }}$ | CLK Pulse Width HIGH |  | 50.0 |  |  | ns |
| $\mathrm{t}_{\text {CWL }}$ | CLK Pulse Width LOW |  | 50.0 |  |  | ns |
| $\mathrm{t}_{\mathrm{ES}}$ | CLK to LE Set Up Time |  | 50.0 |  |  | ns |
| tEW | LE Pulse Width |  | 50.0 |  |  | ns |



| RFout (MHz) | $\begin{aligned} & \mathrm{R} \\ & \Omega \end{aligned}$ | $\begin{aligned} & j X \\ & \Omega \end{aligned}$ | $\left\lvert\, \begin{gathered} \|R+j X\| \\ \Omega \end{gathered}\right.$ |
| :---: | :---: | :---: | :---: |
| 2087.73 | 26.406 | -34.650 | 46.564 |
| 2105.28 | 25.385 | -30.800 | 39.913 |
| 2121.00 | 23.898 | -28.122 | 36.905 |
| 2155.14 | 19.979 | -23.102 | 30.543 |


| Fin | $R$ | $j X$ | $\|R+j X\|$ |
| :---: | :---: | :---: | :---: |
| $(\mathrm{MHz})$ | $\Omega$ | $\Omega$ | $\Omega$ |
| 170.76 | 33.789 | -239.220 | 241.595 |
| 367.20 | 26.992 | -137.620 | 140.242 |
| 440.76 | 27.844 | -126.470 | 129.499 |

## Serial Data Input Timing



Figure 1. Serial Data Input Timing

## NOTE

1. DATA is clocked into the 24 -bit shift register on the rising edge of CLK.
2. The MSB of DATA is shifted in first.

## FUNCTIONAL DESCRIPTION

LMX2542 is a highly integrated, high performance, low power, frequency synthesizer system optimized for Cellular-CDMA 1xRTT and IS-95 mobile handsets and data systems with GPS capabilities. Using a proprietary digital phase locked loop technique, LMX2542 generates very stable, low noise local oscillator (LO) signals for up and down conversion in wireless communications devices.
LMX2542 includes a Voltage Controlled Oscillator (VCO) for the Cellular-CDMA and GPS frequency bands, a loop filter, and a Fractional-N RF PLL based on a $\Delta \Sigma$ modulator which supports frequency resolutions as low as 5 kHz . In concert, these blocks form a closed loop RF synthesizer system. The RF synthesizer system operates from 2087.73 MHz to 2155.14 MHz . The need for external components is limited to a few passive elements for matching the RF output impedance, and bypass elements for power line stabilization.
The Fractional-N RF PLL ( $\Delta \Sigma$ modulator architecture) delivers low spurious thus providing a significant improvement over other PLL solutions. In addition, the Fractional-N RF PLL facilitates faster lock times, which reduces power consumption and system set-up time. Furthermore, the RF loop filter occupies a much smaller area as opposed to the Integer- N architecture. This allows the RF loop filter to be embedded into the circuit, thus minimizing the external noise coupling.
LMX2542 includes an Integer-N IF PLL also. For more flexible loop filter designs, the IF PLL includes a 4 -level programmable charge pump. Together with an external VCO and loop filter, LMX2542 makes a complete closed loop IF synthesizer system. The default IF frequency is 367.20 MHz .
The circuit also supports commonly used reference oscillator frequencies of 19.20 MHz and 19.68 MHz .

## FREQUENCY GENERATION

## RF Frequency Selection

The RF synthesizer (Cellular-CDMA) divide ratio can be calculated using the following equation:

$$
f_{R F o u t}=\left(8 \cdot R F_{-} B+R F_{-} A+\frac{10^{4} \cdot R F_{-} F N}{f_{\text {OSCin }}}\right) \cdot f_{O S C i n}
$$

where

- RF_A < RF_B
- $\mathrm{f}_{\text {RFout }}$ : RF VCO output frequency
- $\mathrm{f}_{\text {oscin }}$ : Reference oscillator frequency
- RF_A : Preset divide ratio of the RF PLL binary 3-bit swallow counter ( $0 \leq \operatorname{RF}$ _A $\leq 7$ )
- RF_B : Preset divide ratio of the RF PLL binary 4-bit programmable counter ( $2 \leq R F \_B \leq 15$ )
- RF_FN : Preset numerator of the RF PLL binary 11 -bit modulus counter ( $0 \leq$ RF_FN $<1920$ for foscin $=19.20 \mathrm{MHz}$ ) ( $0 \leq$ RF_FN < 1968 for $f_{\text {oscin }}=19.68 \mathrm{MHz}$ )


## NOTE

When the FREQ_OFF bit is set to 1 , frequencies with 5 kHz resolution can be generated. In the same way outlined above, the divide ratio for the desired frequency less 5 kHz should be programmed. When the FREQ_OFF bit (R1[2]) is set to 1, the programmed frequency will be shifted by +5 kHz in order to achieve the desired frequency. Refer to RF Synthesizer System Frequency Offset for details on how to program the FREQ_OFF bit.

## IF Frequency Selection

The IF synthesizer divide ratio can be calculated using the following equation:
where

- IF_A < IF_B
- $\mathrm{f}_{\text {Fin }}$ : IF VCO output frequency
- $\mathrm{f}_{\text {oscin }}$ : Reference oscillator frequency
- IF_A : Preset divide ratio of the IF PLL binary 4-bit swallow counter ( $0 \leq \mathrm{IF}$ _A $\leq 15$ )
- IF_B : Preset divide ratio of the IF PLL binary 9-bit programmable counter ( $1 \leq \mathrm{IF}$ _ $\mathrm{B} \leq 511$ )
- IF_R : Preset divide ratio of the IF PLL binary 9-bit programmable reference counter ( $2 \leq \mathrm{IF} \_\mathrm{R} \leq 511$ )

From the above equation and with the SPI_DEF bit set to 1, LMX2542 generates a fixed IF frequency of 367.20 MHz as follows:

| $\mathbf{f}_{\text {Fin }}$ <br> $(\mathbf{M H z})$ | IF_B | IF_A | $\mathbf{f}_{\text {OSCin }} / \mathbf{I F} \mathbf{( k H z}$ <br> $(\mathbf{k H z})$ |
| :---: | :---: | :---: | :---: |
| 367.20 | 191 | 4 | 120 |

## VCO FREQUENCY TUNING

The center frequency of the RF VCO is determined mainly by the resonant frequency of the tank circuit. This tank circuit is implemented on-chip and requires no external inductor. LMX2542 actively tunes the tank circuit to the required frequency with the built-in tracking algorithm.

## POWER CONTROL

LMX2542 includes a powerdown mode to reduce the power consumption. LMX2542 can be powered down when the CE pin is set LOW, independent of the state of the powerdown bits. When CE is set HIGH, powerdown is controlled through the MICROWIRE. The RF and IF circuitries are individually powered down by setting the RF_EN (R1[3]) and IF_EN (R2[2]) bits LOW respectively. Refer to RF Synthesiser System Enable and IF Synthesizer Enable for details on how to program the RF_EN and IF_EN bits.

| CE Pin | RF_EN | IF_EN | RF Circuitry | IF Circuitry |
| :---: | :---: | :---: | :---: | :---: |
| 0 | X | X | OFF | OFF |
| 1 | 0 | 0 | OFF | OFF |
| 1 | 0 | 1 | OFF | ON |
| 1 | 1 | 0 | ON | OFF |
| 1 | 1 | 1 | ON | ON |

1. X refers to a don't care condition.
2. The RF circuitry includes the whole RF synthesizer system (synthesizer and VCO).
3. The IF circuitry includes the IF synthesizer block only.

## RF DIGITAL FILTERED LOCK DETECT

A digital filtered lock detect status generated from the RF phase frequency detector (PFD) is available on the LD pin (Pin 19) when the RF_LD bit (R0[21]) is set to 1. The LD output is therefore used to indicate the lock status of the RF synthesizer system. Furthermore, the LD output can be forced to GND at all times when the RF_LD bit is set to 0 .

When used as a lock detect output, the two inputs to the PFD, $f_{N}$ and $f_{R}$, are first divided by 64 . The lock detect digital filter then compares the difference between the phases of the inputs to the PFD to an RC generated delay of approximately 10 ns. This delay is represented by $\mathrm{t}_{\mathrm{w}}$ in Figure 2 and Figure 3 below. If the phase error is less than $10 \mathrm{~ns}\left(\Delta \mathrm{t}<\mathrm{t}_{\mathrm{w}}\right)$ for 4 consecutive PFD comparison cycles, the RF PLL enters a locked state and the LD output is then forced HIGH. Once the phase error becomes greater than $10 \mathrm{~ns}\left(\Delta t>t_{w}\right)$ the RF PLL falls out of lock and the LD is forced LOW ( $\sim$ GND). The phase error in Figure 3 is measured on the leading edge. If the phase difference between the two inputs to the PFD is equal to $10 \mathrm{~ns}\left(\Delta t=t_{w}\right.$ ), then the LD output becomes unpredictable. Refer to RF Synthesizer Lock Defect for further details on how to program the digital filtered lock detect.

## NOTE

$f_{R}$ is the PFD input from the reference oscillator and $f_{N}$ is the PFD input from the programmable feedback divider ( N counter).


Figure 2. Lock Detect Flow Diagram


LD

Figure 3. Lock Detect Timing Diagram Waveform

## MICROWIRE SERIAL INTERFACE

The programmable register set is accessed via the MICROWIRE serial interface. The interface comprises three signal pins: CLK, DATA, and LE. Serial data is clocked into the 24 -bit shift register on the rising edge of CLK. The least significant bits decode the internal control register address. When LE transitions from LOW to HIGH, DATA stored in the shift registers is loaded into one of six control registers. The MSB of DATA is loaded in first. The synthesizers can be programmed even in power down mode. A complete programming description is provided in Programming Description.

## Programming Description

## MICROWIRE INTERFACE

The MICROWIRE Serial Port Interface (SPI) has a 24 -bit shift register to store the incoming DATA bits temporarily. The incoming DATA is loaded into the shift register from MSB to LSB. The data is shifted at the rising edge of the CLK signal. When the LE signal transitions from LOW to HIGH, the DATA stored in the shift register is transferred to the proper register depending on the state of the ADDRESS bits. The selection of the particular register is determined by the address bits equal to the binary representation of the number of the control register.
At start-up, the 24 -bit shift register is loaded via the MICROWIRE interface. The loading requires 3 default words, with register R2 loaded first, and R0 loaded last. Once loaded, the RF VCO frequency can then be changed by only programming register R0 appropriately. If an IF frequency other than the default value is desired, the SPI_DEF bit should be set to 0 , and registers R4 and R5 programmed appropriately.

## Control Register Content Map

The control register content map describes how the bits within each control register are allocated to specific control functions. The bits that are marked 0 should be programmed as such to ensure proper device operation.

Table 1. Control Register Content Map


## NOTE

Numbers in Bold represent the ADDRESS bits.

## RO REGISTER

The R0 register contains the RF_FN, RF_A, RF_B, RF_LD, and SPI_DEF control words. The register address bits are $R 0[1: 0]=00$. The detailed descriptions and programming information for each control word is discussed in the following sections.

| Reg | 23 | 2 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB |  |  |  |  |  |  |  |  | SHIFT | REG | STE | BI | LO | TIO |  |  |  |  |  |  |  |  | SB |
|  | DATA[21:0] FIELD |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | ADDRESS <br> [1:0] <br> FIELD |  |
| R0 | $\begin{aligned} & \mathrm{SPI} \\ & \mathrm{DEF} \end{aligned}$ | 1 | RF LD | 0 | $\begin{aligned} & \text { RF_B } \\ & {[3: 0]} \end{aligned}$ |  |  |  | $\begin{aligned} & \text { RF_A } \\ & {[2: 0]} \end{aligned}$ |  |  | $\begin{gathered} \text { RF_FN } \\ {[10: 0]} \end{gathered}$ |  |  |  |  |  |  |  |  |  |  | 0 | 0 |

## RF_FN[10:0] - RF Synthesizer Fractional Numerator Counter (R0[2:12])

The RF_FN control word is used to setup the 11-bit $\Delta \Sigma$ modulator. This corresponds to programming the fractional numerator counter portion of the RF feedback divider. The value programmed is dependent on the reference oscillator used.

## Programming RF_FN[10:0] Using 19.20 MHz Reference Oscillator

When a 19.20 MHz reference oscillator is used (OSC_FREQ bit $=0$ ), the RF_FN can be programmed to values ranging from 0 to 1919.

| Numerator | $\begin{gathered} \text { RF_FN[10:0] } \\ \text { foscin }^{\text {O }} 19.20 \mathrm{MHz} \end{gathered}$ |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| - | - | - | - | - | - | - | - | - | - | - | - |
| 1919 | 1 | 1 | 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

## Programming RF_FN[10:0] Using 19.68 MHz Reference Oscillator

Similarly, when a 19.68 MHz reference oscillator is used (OSC_FREQ bit = 1), the RF_FN can be programmed to values ranging from 0 to 1967.

| Numerator | $\begin{gathered} \text { RF_FN[10:0] } \\ \text { fosCin }^{\text {a }}=19.68 \mathrm{MHz} \end{gathered}$ |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| - | - | - | - | - | - | - | - | - | - | - | - |
| 1967 | 1 | 1 | 1 | 1 | 0 | 1 | 0 | 1 | 1 | 1 | 1 |

## RF_A[2:0] - RF Synthesizer Swallow Counter (A Counter) (R0[13:15])

The RF_A control word is used to setup the RF synthesizer's A counter. The A counter is a 3-bit swallow counter used in the programmable feedback divider. The RF_A control word can be programmed to values ranging from 0 to 7 .

| Divide Ratio | RF_A[2:0] <br> RFMode |  |  |
| :---: | :---: | :---: | :---: |
|  | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 1 |
| $\cdot$ | $\cdot$ | $\cdot$ | $\bullet$ |
| 7 | 1 | 1 | 1 |

## RF_B[3:0] - RF Synthesizer Programmable Binary Counter (B Counter) (R0[16:19])

The RF_B control word is used to setup the RF synthesizer's B counter. The B counter is a 4-bit programmable binary counter used in the programmable feedback divider. The RF_B control word can be programmed to values ranging from 2 to 15 . Divide ratios less than 2 are prohibited.

| Divide Ratio | RF_B[3:0] |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| 2 | 0 | 0 | 1 | 0 |
| 3 | 0 | 0 | 1 | 1 |
| $\cdot$ | $\cdot$ | $\mathbf{~}$ | 1 | $\boldsymbol{1}$ |
| 15 | 1 | 1 | 1 |  |

## RF_LD - RF Synthesizer System Lock Detect (RO[21])

The RF_LD bit is used to indicate the lock status of the RF synthesizer system.

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | 0 | 1 |
| RF_LD | R0[21] | RF Synthesizer System <br> Lock Detect | Hard Zero <br> (GND) | Lock Detect |

## SPI_DEF - Serial Port Interface Default Register Selection (R0[23])

The SPI_DEF bit selects between using the default IF counter values and user programmable values.

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :---: | :--- | :--- |
|  |  | 0 | 1 |  |
| SPI_DEF | R0[23] | Serial Port Interface <br> Default Register Selection | Default Counter Values <br> OFF. <br> Program Registers R0 <br> through R6 | Default Counter Values <br> ON. <br> Program Registers R0 <br> through R2 |

## R1 REGISTER

The R1 register contains the FREQ_OFF, RF_EN and OB_CRL control words. The register address bits are R1[1:0] $=01$. The detailed descriptions and programming information for each control word is discussed in the following sections.

| Reg | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 |  |  |  | 9 | 8 | 7 |  | 5 | 4 | 3 | 2 | $\begin{array}{c\|c} 1 & 0 \\ \hline & \text { LSB } \end{array}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | DATA[21:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | ADDRESS <br> [1:0] <br> FIELD |  |
| R1 | 0 | 1 | 1 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 1 | 0 |  |  |  | 1 | 0 | 0 |  | ${ }_{[1:}^{\mathrm{OB}_{1}}$ |  | $\begin{aligned} & \text { RF } \\ & \text { EN } \end{aligned}$ | $\begin{aligned} & \text { FRE } \\ & Q_{-} \\ & \text {OFF } \end{aligned}$ | 0 | 1 |

## FREQ_OFF - RF Synthesizer System Frequency Offset (R1[2])

The FREQ_OFF bit is used to offset the RF frequency by +5 kHz .

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | 0 | 1 |
| FREQ_OFF | R1[2] | RF Synthesizer System <br> Frequency Offset | RF Synthesizer System <br> Frequency Offset <br> Disabled | RF Synthesizer System <br> Frequency Offset Enabled |

## RF_EN - RF Synthesizer System Enable (R1[3])

The RF_EN bit is used to switch the RF synthesizer system (PLL and VCO) between a powered up and powered down mode.

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :--- | :--- | :--- |
| RF_EN | R1[3] | RF Synthesizer System <br> Enable | RF Synthesizer System <br> Powered Down | RF Synthesizer System <br> Active |

## OB_CRL[1:0] - RF VCO Output Buffer Power Control (R1[5:4])

The OB_CRL word is used to set the RF VCO output buffer power level. The power level can be set according to the system requirements.

| OB_CRL[1:0] |  | RF VCO Output Buffer Power Level <br> $(\mathrm{dBm})$ |
| :---: | :---: | :---: |
| 0 | 0 | -4.5 |
| 0 | 1 | -2.0 |
| 1 | 0 | 0.5 |
| 1 | 1 | 3.0 |

## R2 REGISTER

The R2 register contains the IF_EN, IF_CUR, IF_FREQ, OSC_FREQ, and VCO_CUR control words. The register address bits are R2[1:0] = 10. The detailed descriptions and programming information for each control word is discussed in the following sections.

| Reg | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  | DATA[21:0] FIELD |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | ADDRESS <br> [1:0] <br> FIELD |  |

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| R2 | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | VCO_C <br> $[1: 0]$ | OSC <br> FRE <br> Q | IF_FRE <br> Q <br> $[1: 0]$ | IF_CUR <br> $[1: 0]$ | IF <br> EN | $\mathbf{1}$ | $\mathbf{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

## IF_EN - IF Synthesizer Enable (R2[2])

The IF_EN bit is used to switch the IF synthesizer between a powered up and powered down mode.

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | 0 | 1 |
| IF_EN | R2[2] | IF Synthesizer Enable | IF Synthesizer Powered <br> Down | IF Synthesizer Active |

## IF_CUR[1:0] - IF Synthesizer Charge Pump Current Gain (R2[4:3])

The IF_CUR control word is used to set the IF synthesizer's charge pump current gain. Four gain levels are available.

| IF_CUR[1:0] |  | IF Synthesizer <br> Charge Pump Current Gain <br> $(\mu \mathrm{A})$ |
| :---: | :---: | :---: |
| 0 | 0 | 100 |
| 0 | 1 | 200 |
| 1 | 0 | 300 |
| 1 | 1 | 800 |

## IF_FREQ[1:0] - IF Synthesizer Fixed Frequency Selection (R2[6:5])

The IF_FREQ control word is used to set the default fixed IF frequency applicable to the specific CDMA system. For LMX2542, the default fixed IF frequency is 367.20 MHz .

| IF_FREQ[1:0] |  | Fixed IF Frequency <br> $(\mathrm{MHz})$ |
| :---: | :---: | :---: |
| 0 | 0 | 170.76 |
| 0 | 1 | 367.20 |
| 1 | 0 | 440.76 |

## OSC_FREQ - Reference Oscillator Frequency Select (R2[7])

The OSC_FREQ bit is used to select the appropriate reference oscillator frequency.

| Control Bit | Register Location | Description | Function |  |
| :---: | :---: | :--- | :--- | :--- |
|  |  |  | 0 | 1 |
| OSC_FREQ | R2[7] | Reference Oscillator <br> Select | 19.20 MHz Reference <br> Oscillator Selected | 19.68 MHz Reference <br> Oscillator Selected |

## VCO_CUR[1:0] - RF VCO Dynamic Current (R2[9:8])

The VCO_CUR control word is used to set the dynamic current for the RF VCO. A maximum dynamic current is recommended, and is achieved when VCO_CUR[1:0] word = 11 .

| VCO_CUR[1:0] |  | RF VCO Current Magnitude |
| :---: | :---: | :---: |
| 0 | 0 | Minimum |
| $\bullet$ | $\bullet$ | $\bullet$ |
| $\bullet$ | $\bullet$ | $\bullet$ |
| 1 | 1 | Maximum |

## R3 REGISTER

The R3 register is used for internal testing of the device and is not intended for customer use. The register address bits are R3[2:0] = 011. Register R3 is active only when the SPI_DEF bit in Register R0 is set to 0 .

| Reg | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
|  | DATA[20:0] FIELD |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{gathered} \text { ADDRESS } \\ {[2: 0]} \\ \text { FIELD } \end{gathered}$ |  |  |
| R3 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 |

## R4 REGISTER

The R4 register contains the IF_B and IF_A control words. The register address bits are R4[3:0] = 0111. Register R4 is active only when the SPI_DEF bit in Register R0 is set to 0 . Regsiter R4 should only be used to set the IF N counter if the default value is not desired. The detailed descriptions and programming information for each control word is discussed in the following sections.

| $\begin{gathered} \mathrm{R} \\ \mathrm{eg} \end{gathered}$ | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB |  |  |  |  |  |  |  |  | SHIF | RE | STE | BIT | OC | IO |  |  |  |  |  |  |  |  | LSB |
|  | DATA[19:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | ADDRESS <br> [3:0] <br> FIELD |  |  |  |
| $\begin{aligned} & \mathrm{R} \\ & 4 \end{aligned}$ | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |  |  |  |  |  | $\begin{aligned} & \mathrm{IF}= \\ & {[8: 0} \end{aligned}$ |  |  |  |  | 0 | 1 | 1 | 1 |

## IF_B[8:0] - IF Synthesizer Programmable Binary Counter (B Counter) (R4[12:4])

The IF_B control word is used to setup the IF synthesizer's B counter. The B counter is a 9-bit programmable binary counter used in the programmable feedback divider. The IF_B control word can be programmed to values ranging from 1 to 511 . Divide ratios less than 1 are prohibited.

| Divide Ratio | IF_B[8:0] |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| 2 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| - | - | - | - | - | - | - | - | - | - |
| 511 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

## IF_A[3:0] - IF Synthesizer Swallow Counter (A Counter) (R4[16:13])

The IF_A control word is used to setup the IF synthesizer's A counter. The A counter is a 4-bit swallow counter used in the programmable feedback divider. The IF_A control word can be programmed to values ranging from 0 to 15 .

| Divide Ratio | IF_A[3:0] |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 1 |
| $\cdot$ | $\cdot$ | $\mathbf{1}$ | $\mathbf{0}$ | $\boldsymbol{0}$ |
| 15 | 1 | 1 | 1 |  |

## R5 REGISTER

The R5 register contains the IF_R control word. The register address bits are R5[4:0] = 01111. Register R5 is active only when the SPI_DEF bit in Register R0 is set to 0 . Regsiter R5 should only be used to set the IF R counter if the default value is not desired. The detailed description and programming information for this control word is discussed in the following section.
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| Reg | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MSB | SHIFT REGISTER BIT LOCATION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | LSB |
|  | DATA[18:0] FIELD |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | ADDRESS <br> [4:0] <br> FIELD |  |  |  |  |
| R5 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |  | $\begin{aligned} & \text { IF_R } \\ & {[8: 0]} \end{aligned}$ |  |  |  |  | 0 | 1 | 1 | 1 | 1 |

## IF_R[8:0] - IF Synthesizer Programmable Reference Divider (R5[13:5])

The IF_R control word is used to setup the IF synthesizer's reference divider. The IF_R control word can be programmed to values ranging from 2 to 511 . Divide ratios less than 2 are prohibited.

| Divide Ratio | IF_R[8:0] |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 2 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| 3 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| - | - | - | - | - | - | - | - | - | - |
| 511 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

## R6 REGISTER

The R6 register is used for internal testing of the device and is not intended for customer use. The register address bits are R6[5:0] $=01111$. Register R6 is active only when the SPI_DEF bit in Register R0 is set to 0 .


## REVISION HISTORY

[^1]
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| Products |  | Applications |  |
| :---: | :---: | :---: | :---: |
| Audio | www.ti.com/audio | Automotive and Transportation | www.ti.com/automotive |
| Amplifiers | amplifier.ti.com | Communications and Telecom | www.ti.com/communications |
| Data Converters | dataconverter.ti.com | Computers and Peripherals | www.ti.com/computers |
| DLP® Products | www.dlp.com | Consumer Electronics | www.ti.com/consumer-apps |
| DSP | dsp.ti.com | Energy and Lighting | www.ti.com/energy |
| Clocks and Timers | www.ti.com/clocks | Industrial | www.ti.com/industrial |
| Interface | interface.ti.com | Medical | www.ti.com/medical |
| Logic | logic.ti.com | Security | www.ti.com/security |
| Power Mgmt | power.ti.com | Space, Avionics and Defense | www.ti.com/space-avionics-defense |
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[^0]:    今
    Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of
    Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

[^1]:    - Changed layout of National Data Sheet to TI format23

